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**Circuit arrangement with power factor correction, and  
corresponding appliance**

The present invention relates to a circuit arrangement having a switch-mode power supply as claimed in the preamble clause of claim 1, and on an appliance having a corresponding circuit arrangement.

Switch-mode power supplies produce a heavily pulsed load on the electrical power supply system, which leads to harmonic currents in electrical power supply systems. This load occurs in the region of the voltage maxima of the sinusoidal mains voltage, at which voltage maxima an energy-storage capacitor for the switch-mode power supply is recharged. Appliances having a relatively high power consumption, such as televisions with relatively large picture tubes, now therefore have to comply with specific regulations relating to harmonic currents. The harmonic load on the electrical power supply system caused by an appliance can in this case be stated as a so-called power factor.

Widely differing circuit concepts are known for improving the power factor, for example from DE-A-196 10 762, EP-A-0 700 145 and US 5,986,898. These contain a second current path with a coil between the mains rectifier and the primary winding of the transformer, with the inductance of this coil acting like a current pump, which is controlled by a switching transistor in the switch-mode power supply and in consequence broadens the pulsed current flow of the switch-mode power supply.

A further possible way to improve the power factor of a switch-mode power supply is to use a coil in the input area of the switch-mode power supply. This coil is also referred to as a 50 Hz coil, a mains frequency coil or a power factor coil. In order to avoid confusion with

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other coils, the description therefore always uses the expression power factor coil for this coil.

However, this power factor coil has the disadvantage  
5 that the current flow in the power factor coil is interrupted suddenly when the mains switch is operated in order to switch off the appliance. The energy which is stored in the coil must be dissipated, however. Since the open mains switch represents the highest  
10 impedance in the circuit, a very high voltage is therefore developed across the switching contacts of the mains switch, and causes an arc. This causes the mains switch to age more quickly, and the switch represents a safety risk since, in the worst case, it  
15 becomes a potential fire source.

Other mains switches, in which the switching contacts open at a slow rate, are likewise at risk. In this case, although the voltage across the contacts is not  
20 very high, an arc is nevertheless produced, however, and burns until the end of the corresponding power supply system half-cycle. In this case as well, a large amount of energy is lost in the switch, which leads to rapid aging.

25 The object of the present invention is to specify a circuit arrangement and a corresponding appliance of the type mentioned initially, which have a high reliability, combined with a low circuit complexity.

30 For a circuit arrangement, this object is achieved by the features of claims 1, 7 and 8, and for an appliance it is achieved by the features of claim 9. Advantageous developments of the invention are specified in the  
35 dependent claims.

The circuit arrangement according to the invention has a mains connection, a mains switch with two switching

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contacts, a demagnetization coil and a switch-mode power supply, which comprises a driver circuit for producing a control voltage for the switching transistor of the switch-mode power supply. A first  
5 switching contact of the mains switch is arranged here between the mains connection and the demagnetization coil in order to switch the demagnetization coil on and off, and the second switching contact is connected to a supply or control voltage for the driver circuit, in  
10 order to switch off the control voltage for the switching transistor or in order to switch off the switch-mode power supply.

This has the advantage that a power factor coil between  
15 the mains connection and the switch-mode power supply can be used for power factor correction, and that the mains switch is not loaded by the inductance of the power factor coil. Admittedly, this results in the switch-mode power supply being connected to the mains  
20 even when it is switched off. However, this is not disadvantageous for the user and leads only to a slight power loss resulting from the starting circuit, when the switch-mode power supply is switched off.

25 The circuit arrangement can be used in particular in appliances having a picture tube, for example televisions and computer monitors. This mains switch circuitry in this case ensures that the picture tube is demagnetized whenever the appliance is switched on.

30 The invention will be explained in more detail in the following text using, by way of example, an exemplary embodiment which is illustrated schematically in the figures, in which:

35 Figure 1 shows a circuit arrangement having a mains connection, a power factor coil and a switch-

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mode power supply, as well as a mains switch,  
and

Figure 2 shows the use of the arrangement in a chassis  
of a television.

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Figure 1 shows a switch-mode power supply I  
schematically, which has a rectifier means, in this  
exemplary embodiment a bridge rectifier BR with four  
diodes, an energy-storage capacitor C1 and transformer  
10 TR with a primary winding W1, an auxiliary winding W2  
arranged on the primary side and secondary windings W3,  
W4 and W5 arranged on the secondary side. The energy-  
storage capacitor C1 is in this case arranged between  
the bridge rectifier BR and the primary winding W1. A  
15 switching transistor T1 is connected in series with the  
primary winding W1 and is controlled by a control  
voltage DS for a driver circuit DC. A supply voltage  
VCC is provided for operation of the switch-mode power  
supply via the auxiliary winding W2, a diode D1 and  
20 capacitor C2, and is supplied to the driver circuit DC.

The switch-mode power supply I in Figure 1 is a part of  
a circuit arrangement which also has a mains connection  
NA, a demagnetization coil (which is not illustrated)  
25 and a mains switch S1 with two switching contacts 1, 2.  
The circuit arrangement is preferably integrated in an  
appliance, for example a television, which has a  
picture tube.

30 Picture tubes, which are used in televisions or  
computer monitors, need to be demagnetized from time to  
time in order to preserve the color purity of the  
picture tube. This is accomplished by means of a  
demagnetization coil, through which an AC voltage  
35 normally flows during the process of switching on the  
appliance. The 220 volt mains voltage is used as the AC  
voltage in this case and produces a high current surge  
at the time when the appliance is switched on, and

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which then subsequently decays gradually. The decay is accomplished by means of a posistor, which is heated by the high current with its impedance in consequence becoming high.

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In order to regulate the switch-mode power supply, the driver circuit DC is supplied with a regulating signal RS which is derived from a supply voltage U4 on the secondary side, for example from the system voltage in a television, and is transmitted via an opto coupler or an isolating transformer (which is not illustrated) to the primary side of the switch-mode power supply.

15 A so-called snubber network SN is arranged in parallel with the primary winding W1 and is used to damp voltage spikes which occur during the process of switching off the switching transistor T1. The switch-mode power supply also contains a starting circuit AS, which supplies the driver stage DC with a current for the starting phase of the switch-mode power supply after the appliance in which the circuit arrangement is integrated has been switched on. The starting circuit AS is normally a high-value resistor chain, which produces a connection between the bridge rectifier BR and the capacitor C2. During operation, the supply voltage VCC is produced by the auxiliary winding W2, as well as by the diode D1 and the filter capacitor C2.

30 The switch-mode power supply illustrated in the figure preferably operates on the basis of the fly back converter principle, although other circuit principles are likewise possible. Fly back converters are preferably used in entertainment electronics appliances, for example in televisions and video recorders. In a fly back converter, energy is in this case stored in the transformer TR during the phase in which the switching transistor T1 is switched on, and is then subsequently transmitted to the windings W3-W5

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on the secondary side, and to the auxiliary winding W2 on the primary side, in the phase during which the switching transistor is switched off. Fly back converters are used both as AC/DC converters and as  
5 DC/DC converters.

Switch-mode power supplies of this type have a low power factor, since the energy-storage capacitor C1 is recharged only in the region of the voltage maxima and  
10 minima of the 50 Hz mains voltage, when the output voltage U2 of the bridge rectifier BR exceeds the voltage value across the energy-storage capacitor C1. One simple possible way to improve the power factor of a switch-mode power supply is to use a mains frequency  
15 coil or power factor coil NS, which is arranged between the mains connection NA and the energy-storage capacitor C1. In this exemplary embodiment, it is connected between the mains connection NA and the mains rectifier BR.

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The power factor coil NS results in the pulsed current flow which is used to recharge the energy-storage capacitor C1 being broadened and shifted in phase, since the inductance of this coil, for example 50 mH,  
25 means that the current flow through the coil rises only gradually, and decays again in a damped manner. A power factor which complies with the requirements is achieved in this case by appropriate choice of the inductance value.

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Figure 1 likewise shows a mains switch S1, which has two switching contacts 1 and 2, as well as corresponding connections a and b for the switching contact 1, and c and d for the switching contact 2. The  
35 mains switch S1 is in this case normally arranged on the front face of the appliance, such that it is easily accessible for a user. The connections c and d are in this case according to the invention connected to the

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connections c' and d' for switching the driver circuit DC of the switch-mode power supply on and off, and the connections a and b are connected to the magnetization coil for switching the demagnetization coil on and off.

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In consequence, the power factor coil NS is not arranged in the current path of the mains switch S1. The mains switch is therefore not loaded by the power factor coil NS when the appliance is switched on and off. At the same time, this arrangement ensures that the demagnetization coil is switched on, in order to demagnetize the picture tube, on each occasion when the appliance is switched on.

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15 Figure 2 shows the major components of the circuit arrangement, to the extent that they are integrated on a chassis of a television in an advantageous refinement. On the input side, the arrangement has a filter section with a mains filter NF, capacitors C3 and C4 and a resistor R1, which filter section is connected to the mains connection NA of the arrangement. This filter section prevents radiated interference from the appliance entering the mains system and suppresses any interference which is already present on the mains system. On the output side, the filter section is connected to the bridge rectifier BR, which provides the rectified voltage U2 for operation of the switch-mode power supply.

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30 Connections e and f are arranged in the connection between the bridge rectifier and the filter section, to which the power factor coil NS, which is not illustrated in Figure 2, is connected. The power factor coil is arranged separately, so that the switch-mode power supply can be kept compact and, in particular, such that an already existing switch-mode power supply which does not have power factor correction can be used without any major design change. The arrangement of the

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power factor coil NS upstream of the bridge rectifier BR means that it is located in the current path of the sinusoidal AC voltage of the mains system.

5 The driver circuit DC and the transformer TR for the switch-mode power supply are indicated only schematically in Figure 2, to the extent that it is necessary for understanding of the invention. As already explained with reference to Figure 1, a diode  
10 D1 and a capacitor C2 are arranged on the auxiliary winding W2, which is arranged on the primary side, in order to produce the operating voltage VCC for the driver circuit. Two connection points c' and d' are arranged on the chassis in the connection between the  
15 capacitor C2 and the driver circuit DC, and are connected to the connection points c and d of the mains switch S1.

A capacitor C6 with a capacitance of 1  $\mu$ F is also  
20 advantageously connected to ground between the connection c' and the driver circuit DS, and is used to avoid switch bouncing of the switching contact 2 and to provide filtering for the long supply line to the mains switch S1. The capacitance of this capacitor in this  
25 case also influences the number of switching cycles after which the switching transistor T1 is completely switched off.

A resistor PS and a capacitor C5 to which the  
30 demagnetization coil ES (which is not illustrated) is connected via connection points g and h are also arranged on the chassis. On the input side, the resistor PS is arranged in parallel with the capacitor C3. Connection points a' and b', which are connected to  
35 the connection points a and b of the mains switch S1, are in this case located in a supply. In consequence, the demagnetization coil ES is switched on whenever the mains switch S1 is operated in order to switch on the



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appliance. The posistor PS in this case ensures a large current surge at the moment of switching on, which is reduced continuously to a low loss level by the heating of the posistor.

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Whenever the appliance is switched, when the mains switch S1 is pressed, the switching contacts 1 and 2 are closed, so that the switch-mode power supply can be started via the switching contact 2, since the driver  
10 circuit DC is in this case supplied with the operating voltage VCC. On being switched on, the switch-mode power supply thus behaves in precisely the same way as when using the mains switch S1 in appliances of a conventional type.

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Whenever the appliance is switched off, when the mains switch S1 is opened, the switching contact 2 switches off the switching transistor T1 in a short time, so that no more energy is transmitted in the transformer  
20 TR to the second windings W2 - W5. The switching voltage DS for the switching transistor T1 is switched off directly or indirectly by the driver circuit. The switching contact 2 is advantageously connected to the connection points c' and d', as described with  
25 reference to Figure 2, so that the driver circuit DC is disconnected from the supply voltage VCC during the switching-off process. However, another voltage of the driver circuit DC, for example a control voltage for the driver circuit DC, can also be switched off in a  
30 corresponding manner by the switching contact 2, or the switching contact 2 can be used to regulate the regulating signal RS at a predetermined voltage level, so that the switching transistor T1 is likewise switched off permanently.

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According to the invention, the mains connection NA is thus connected to the bridge rectifier BR without being switched, without a mains switch. The switch-mode power

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supply I is thus always connected to the electrical power supply system whenever the mains plug of the appliance is connected to the electrical power supply system. In consequence, the power factor coil NS is not  
5 arranged in the current path of the mains switch S1, so that the life of the mains switch S1 is considerably improved.

Further refinements of the invention are within the  
10 scope of a person skilled in the art. The invention is not restricted to switch-mode power supplies based on the flyback converter principle, as already explained above, and can also be used for other switch-mode power  
15 supply concepts when power factor correction is required. The invention can also be used in appliances which do not have power factor correction. Thus, for example, the connection points e and f can be short-circuited in appliances for countries which do not  
20 require power factor correction. In consequence, there is no need to modify the chassis in these appliances for these countries.